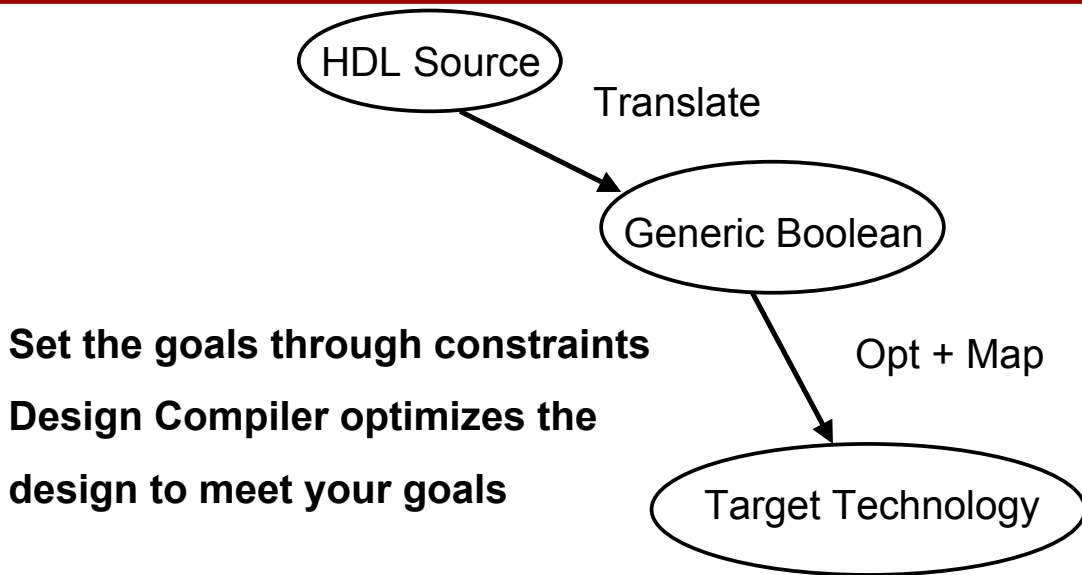


## Synthesis is ...

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## Other information needed for logic synthesis

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- What else does the synthesis tool need besides RTL?
- Library of cells into which to map
- Information about clocks
- Information about inputs
- Information about outputs
- Hints for synthesis tools

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# Target Standard Cell Library

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- Used by Design Compiler for building a circuit
- During mapping, DC will
  - Choose functionally-correct gates from this lib
  - “time” the circuit using supplied timing data for these gates
- Produced either by fab or third parties
  - Often available for “free” (fab adds fee to wafer cost)
- Large set of combinational cells
  - Primitive: INV,NAND, NOR, AOI, OAI, XOR, MUX, etc.
  - Compound: AND,OR, ADD, etc.
  - Sequential, flip-flop/latch, {positive, negative} D-type Flip-flop with {set, reset}

## Needed for each Cell

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- Logical: relationship between outputs and inputs
- Timing/Electrical:
  - Capacitance information for each pin
  - Delay information for each input->output path
  - Slope of outputs
  - Power consumption of cell
  - Required setup time for sequential cells
  - Models are generally tables
- Physical:
  - Layout
  - Abutment and Pin location for place and rout
  - Area of cell for synthesis

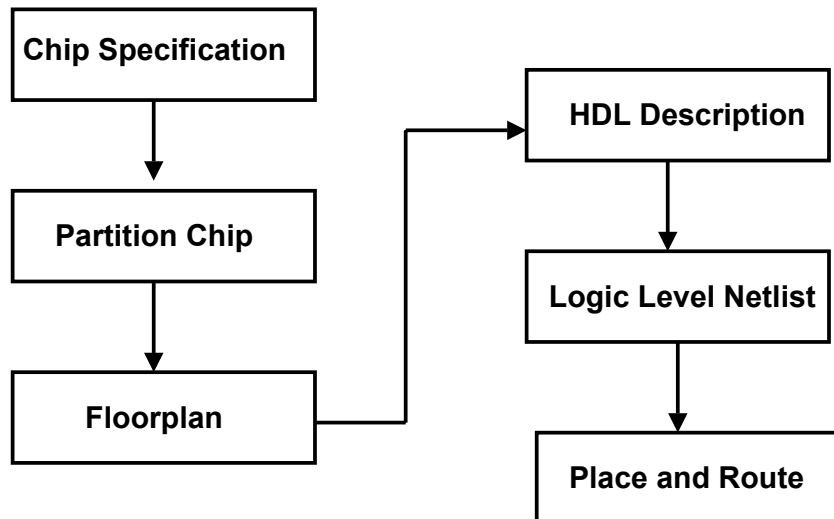
# Partitioning for Synthesis

- Why?
  - Separate distinct functions
  - Workable size and complexity
  - Design reuse
  - Meet physical constraints
  - Good for team projects
- Bad partitioning may end up with long wires and slow design, no opt cross the boundaries
- Related CL and destination registers are grouped into one block
- Balance block size with run time
- Separate core logic, pads, clocks, ...

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## Chip Synthesis Process

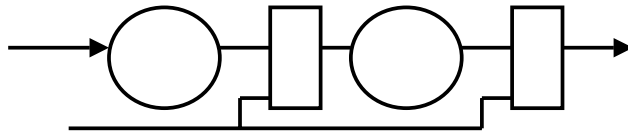


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# Coding for Synthesis

- Think Hardware!
  - Think of topology implied by the code
- Think RTL! (Register Transfer Level)
- Writing in an RTL coding style means describing
  - The register architecture
  - The circuit topology
  - The functionality between registers
- Design Compiler optimizes the logic in between not the register placement



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## Logic Synthesis Example – 4-bit Gray Counter

```
module gray(clk, reset,out);  
  
  input clk, reset;  
  output [3:0] out;  
  
  wire clk,reset;  
  
  reg [3:0] out;  
  
  always @(posedge clk)  
  begin  
    if(reset == 1) out = 4'b0000;  
    else begin  
      case(out)  
        4'b0000: out = 4'b0001;  
        4'b0001: out = 4'b0011;  
        4'b0010: out = 4'b0110;  
        4'b0011: out = 4'b0010;  
        4'b0100: out = 4'b1100;  
        4'b0101: out = 4'b0100;  
        4'b0110: out = 4'b0111;  
        4'b0111: out = 4'b0101;  
        4'b1000: out = 4'b0000;  
        4'b1001: out = 4'b1000;  
        4'b1010: out = 4'b1011;  
        4'b1011: out = 4'b1001;  
        4'b1100: out = 4'b1101;  
        4'b1101: out = 4'b1111;  
        4'b1110: out = 4'b1010;  
        4'b1111: out = 4'b1110;  
      endcase  
    end  
  end  
endmodule
```

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# Synthesis of Gray Counter

- Use 0.13 $\mu$  library
- Set clock period to 2ns
- Set clock skew to 100ps
- Set input drive equal to INVX1
- Set input arrival to be 1ns after clk
- Set output load to be 4 NAND2X2 gates
- Require output valid by 1ns after clock

```
module gray(clk, reset, out);
input clk, reset;
output [3:0] out;

wire clk, reset;

reg [3:0] out;

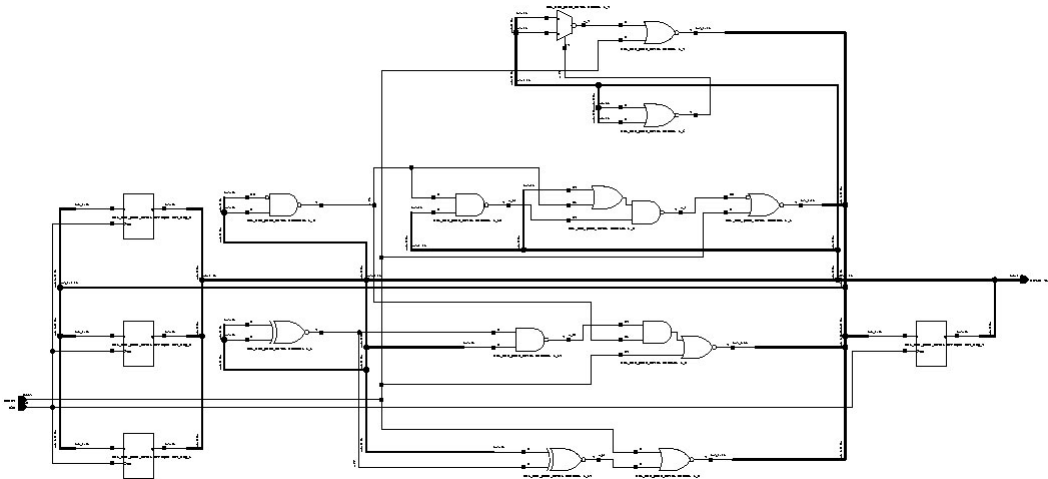
always @(posedge clk)
begin
if(reset == 1) out = 4'b0000;
else begin
case(out)
4'b0000: out = 4'b0001;
4'b0001: out = 4'b0011;
4'b0010: out = 4'b0110;
4'b0011: out = 4'b0010;
4'b0100: out = 4'b1100;
4'b0101: out = 4'b0100;
4'b0110: out = 4'b0111;
4'b0111: out = 4'b0101;
4'b1000: out = 4'b0000;
4'b1001: out = 4'b1000;
4'b1010: out = 4'b1011;
4'b1011: out = 4'b1001;
4'b1100: out = 4'b1101;
4'b1101: out = 4'b1111;
4'b1110: out = 4'b1010;
4'b1111: out = 4'b1110;
endcase
end
end
endmodule
```

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## Results of Synthesis



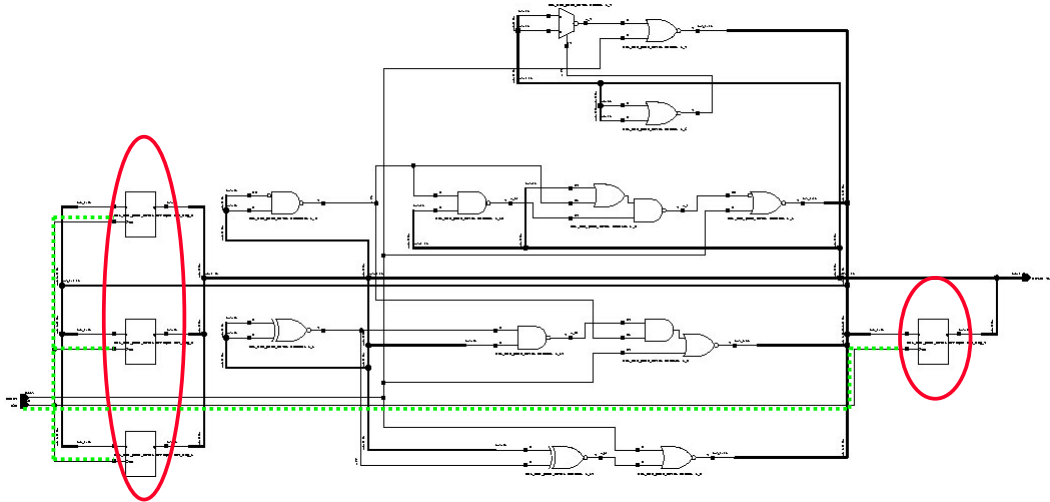
- Read back into schematic system after synthesis
- Gate placement in schematic is somewhat random

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# Results of Synthesis – State and Clocking

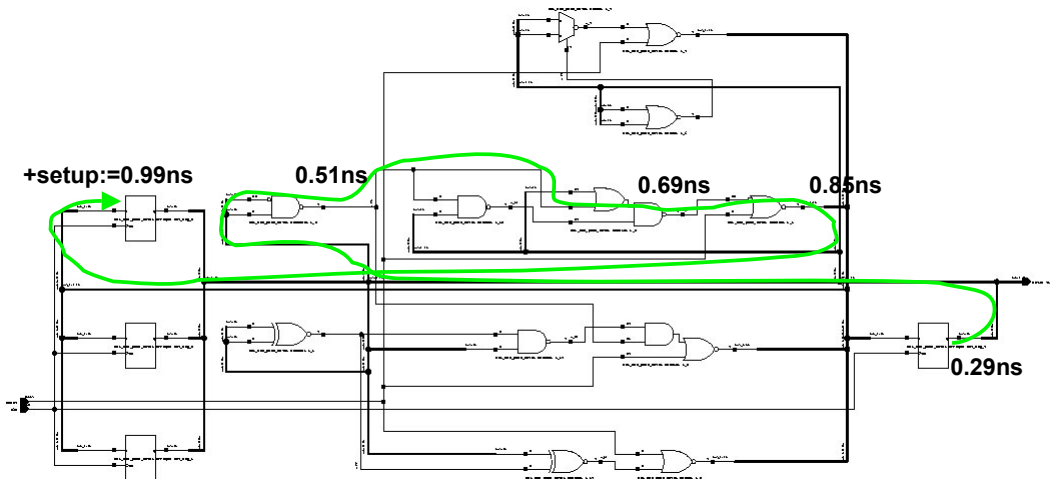


- 4-FF's – corresponds to original code
- All connected to global clock

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# Results of Synthesis – Critical path



- Tool chose gates to meet required cycle time (2ns)
- Plenty of margin

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# Code Style and Synthesis Results

- Suppose we make the counter this way:

```
module gray(clk, reset,out);
input clk, reset;
output [3:0] out;

wire clk,reset;

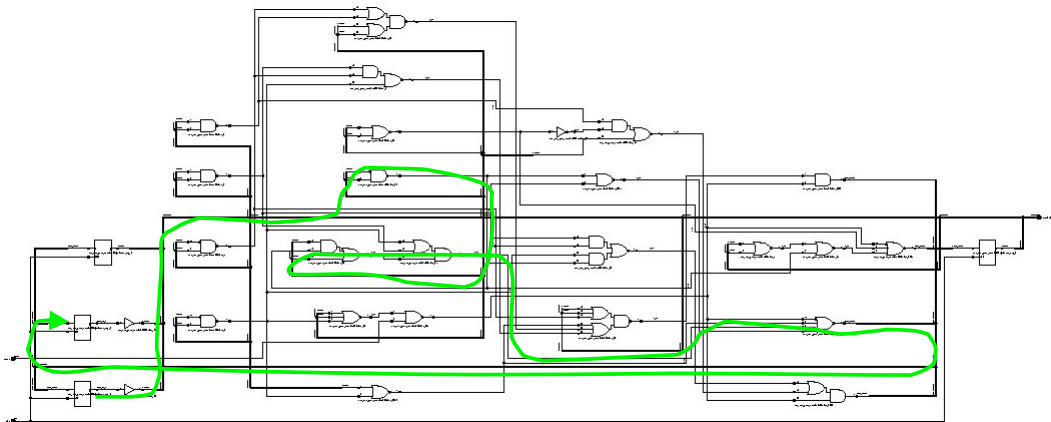
reg [3:0] out;

always @(posedge clk)
begin
    if((reset == 1) || (out == 4'b1000)) out = 4'b0000;
    else if (out == 4'b0000) out = 4'b0001;
    else if (out == 4'b0001) out = 4'b0011;
    else if (out == 4'b0010) out = 4'b0110;
    else if (out == 4'b0011) out = 4'b0010;
    else if (out == 4'b0100) out = 4'b1100;
    else if (out == 4'b0101) out = 4'b0100;
    else if (out == 4'b0110) out = 4'b0111;
    else if (out == 4'b0111) out = 4'b0101;
    else if (out == 4'b1001) out = 4'b1000;
    else if (out == 4'b1010) out = 4'b1011;
    else if (out == 4'b1011) out = 4'b1001;
    else if (out == 4'b1100) out = 4'b1101;
    else if (out == 4'b1101) out = 4'b1111;
    else if (out == 4'b1110) out = 4'b1010;
    else if (out == 4'b1111) out = 4'b1110;
end
endmodule
```

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## If-else Gray Counter



- 26 gates instead of 12
- Slowest path is 1.4ns instead of 1ns

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## *if-else* Statements and case Statements

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- *If-else* Implies multiplexing hardware
- To infer latches, use *if* without an *else* clause
- *If-then-elseif* statements **imply priority**
  - Use only if priority checking is required
  - Otherwise result will be incorrect or possibly slower
- *case* Statements imply parallel mux function
  - Use case statements where possible, particularly for FSM's

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## Combinational Adder Contents

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- 369 cell instances from 1 line of verilog
- Did you really need it?



- Tool created 32-bit Carry Lookahead Adder
  - Tool contains parameterizable set of arithmetic operators
  - Grabs the appropriate one for your problem

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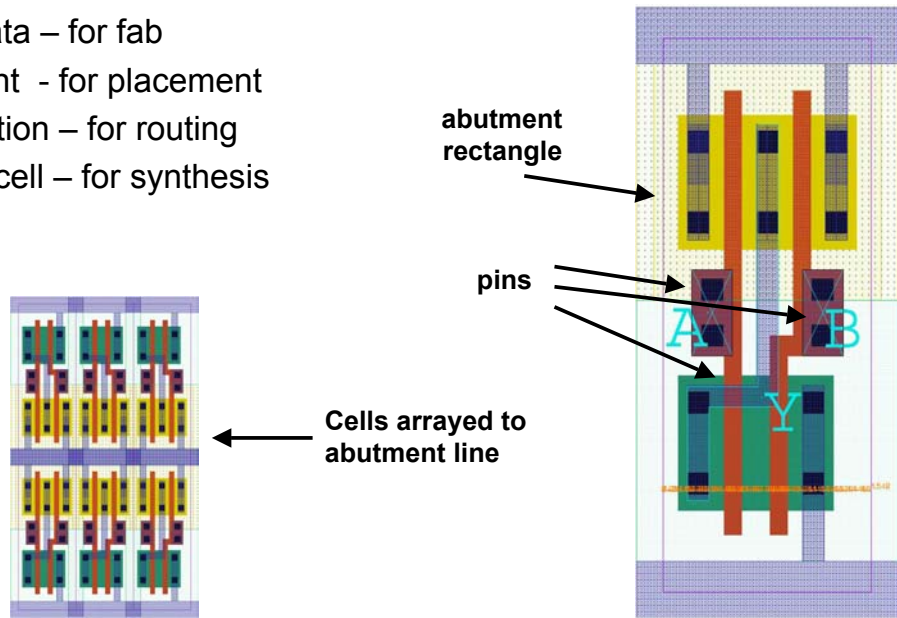
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# Layout for Std Cells

- Mask data – for fab
- Abutment - for placement
- Pin location – for routing
- Area of cell – for synthesis



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## 2-Layer metal Place and Route (Old Style)

- Arrange cells in rows interspersed with routing channels
- Routing runs vertically in m1, horizontally in m2
- No routing over cells

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# Placement for Multiple Levels of Routing

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- No need to leave channels
- Most (80-90%) of surface can be covered by cells